# Continuous Parameter Estimation from Compressive Samples

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# Compressed sensing

• Linear measurements of  $\boldsymbol{s} \in \mathbb{C}^N$ 



- *K*-sparse signal
  - $egin{aligned} oldsymbol{s} &= oldsymbol{\Psi} oldsymbol{lpha} \ &\in \mathbb{C}^N \ &\in \mathbb{C}^N \ &\#\{j: lpha_j 
    eq 0\} \ \|oldsymbol{lpha}\|_0 \leq K \ll N \end{aligned}$

To obtain s, compressive sensing advocates solving

$$\widehat{\boldsymbol{\alpha}} := \underset{\widetilde{\boldsymbol{\alpha}} \in \mathbb{C}^N}{\arg \min} \| \widetilde{\boldsymbol{\alpha}} \|_1 \text{ subject to } \| \boldsymbol{y} - \boldsymbol{\Phi} \boldsymbol{\Psi} \widetilde{\boldsymbol{\alpha}} \|_2 \le \epsilon.$$
 (P1)

• Succeeds, if M is 'large' enough, and  $\Phi \Psi$  is nice

# Useful information without reconstruction

- Sometimes, it is sufficient if we can extract specific information about  $\boldsymbol{s}$
- Example Localisation of *known* patterns in an image



- Biological applications
- 1D: speech/audio processing
- Location is a continuous parameter
- Only compressive samples are available

# Signal model

• Continuous signal model

$$s_c(t) = \sum_{j=1}^{K} \alpha_j f_c(t - \tau_j)$$

•  $\{\alpha_j\}, \{\tau_j\}$  :unknown amplitude and delay



• Fixed sampling grid

• 
$$f_{\tau_j} \in \mathbb{R}^N$$
: vector of samples of  $f_c(t - \tau_j)$   
•  $s = \sum_{j=1}^K \alpha_j f_{\tau_j}$ : discrete signal •  $\{\alpha_j\}, \{\tau_j\}$ : continuous

#### Discrete finite signal dictionary

• Suppose  $K \ll N$ ,

 $m{s}$  is sparse w.r.t a dictionary containing by ALL admissible translations of  $f_c(t)$ 

#### But, impossible!

• So, settle for a finite equivalent

given a finite sampling of delay parameters  $\{\delta_i\}_{i=1}^D$ , with step size  $\Delta$ , build

$$\boldsymbol{\Psi} = \left[ \boldsymbol{f}_{\delta_1}, \boldsymbol{f}_{\delta_2} \dots \boldsymbol{f}_{\delta_D} 
ight]$$

• Small  $\Delta \rightarrow$  strongly correlated columns

• Solving (P1) can only estimate delays with error in  $[-\Delta/2, \Delta/2]$ 

# From on grid to off grid delays

- Let  $\mathcal{M}$  : Set of all delayed and scaled signals
- Columns of  $\Psi$  : linear subspace approximation of  $\mathcal M$





Taylor interpolation around each column

- Idea: Include interpolator functions in  $\Psi$
- Solve specialised  $(\mathbf{P1})$  to obtain the off grid delays

Figures taken from the reference [Ekanadham 2011]

#### Polar interpolation 1/2

- If  $f_c(t)$  is such that
  - $||f_c(t-\delta_i)||_2 = ||f_c(t-\delta_j)||_2$  for all  $\delta_i, \delta_j$ : norm preservation

• 
$$||f_c(t-\delta_i) - f_c(t-\delta_i+\delta)||_2 = ||f_c(t-\delta_i) - f_c(t-\delta_i-\delta)||_2$$

symmetric curvature

- Polar interpolation: Approximation by an arc of a circle
  - Any segment of  $\mathcal{M}$  around  $\delta_i$  can be written as

$$f_c(t-\tau) \approx c(t) + r \cos\left(\frac{2\tau}{\Delta}\theta\right) u(t) + r \sin\left(\frac{2\tau}{\Delta}\theta\right) v(t) \quad \tau \in \left[-\Delta/2, \Delta/2\right]$$

$$\begin{pmatrix} f_c(t-\delta_i+\Delta/2)\\f_c(t-\delta_i)\\f_c(t-\delta_i-\Delta/2) \end{pmatrix} = \begin{pmatrix} 1 & r\cos(\theta) & -r\sin(\theta)\\1 & r & 0\\1 & r\cos(\theta) & r\sin(\theta) \end{pmatrix} \begin{pmatrix} c(t)\\u(t)\\v(t) \end{pmatrix}$$
$$r = \|f_c(t)\|_2 \quad \theta = \text{angle}(f_c(t-\delta_i), f_c(t-\delta_i+\Delta/2))$$

# Polar interpolation 2/2



- Incorporate all the interpolating functions in the dictionary  $\Psi$
- Solve specialised (P1) to obtain the off grid delays



- Perform reconstruction oblivious of delay-scale model
  - Estimate delays in signal domain (matched filtering, etc.)

#### From proxy delays to actual delays

- Idea: Greedy approach
  - $\boldsymbol{y}_{res} = \boldsymbol{y}$
  - Select a column of  $\Psi$  that maximally correlates with residue
  - Use the obtained proxy delay  $\delta_p$  and perform polar interpolation to obtain  $\hat{ au}_j$

$$\boldsymbol{y}_{res} \approx \boldsymbol{\Phi} \begin{bmatrix} \boldsymbol{f}_{\delta_p - \Delta/2} & \boldsymbol{f}_{\delta_p} & \boldsymbol{f}_{\delta_p + \Delta/2} \end{bmatrix} \begin{pmatrix} \begin{bmatrix} 1 & r\cos(\theta) & -r\sin(\theta) \\ 1 & r & 0 \\ 1 & r\cos(\theta) & r\sin(\theta) \end{bmatrix}^{-1} \end{pmatrix}^T \boldsymbol{x}$$

$$\boldsymbol{x} = \begin{bmatrix} \alpha_j \\ \alpha_j r \cos\left(\frac{2\tau\theta}{\Delta}\right) \\ \alpha_j r \sin\left(\frac{2\tau\theta}{\Delta}\right) \end{bmatrix} \qquad \text{solved as least} \qquad \hat{\tau}_j = \delta_j + \arctan\left(\frac{\hat{x}_3}{\hat{x}_2}\right) \frac{\Delta}{2\theta}$$

- Obtain new residue incorporating the interpolated function
- Repeat K times

### Example: Compressive deflectometry 1/2

- Several images  $\{\boldsymbol{s}_k\}$
- Each  $\boldsymbol{s}_k$  characterised by  $\widehat{\tau}_k = (\overline{c}_k^x, \overline{c}_k^y)^T$ .
- $\boldsymbol{f}^{
  ho}_{ au}$ : 2D Gaussian



with radius ho, *translated* by T.



Location of dominant deflection.

• To obtain proxy translation parameter

• Matched filtering 
$$\widehat{ au}_k = rgmax_{ au} |\langle m{s}_k, m{f}^{
ho}_{ au} 
angle|.$$

• Efficiently implemented as convolution

### Example: Compressive deflectometry 2/2

• Compressive matched filtering (Smashed filtering)

$$oldsymbol{y}_k \in \mathbb{R}^M, \widetilde{ au}_k = rgmax_ au |\langle oldsymbol{\Phi}^T oldsymbol{y}_k, oldsymbol{f}_ au^
ho 
angle|.$$

- Efficiently implemented as convolution
- Large enough M
  - achieves maximum at same location as in signal domain

• Translation estimation = 1 convolution + 1 least squares

### Experimental results

- Synthetic
  - randomly generated Gaussian patterns
  - ground truth known
- Experimental
  - deflectometric data of two lenses
  - reference translations: obtain using full reconstruction
- Error: absolute pixel error between estimate and reference



# Summary

- A signal model and method to estimate continuous parameters from compressive measurements
- Possible interesting direction:
  - Learn the function  $f_c(t)$
  - others?
- References
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